



# Adaptive Synchronization for Hyperchaotic Liu System

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In this paper, the adaptive control design is investigated for the chaos synchronization of two identical hyperchaotic Liu systems. First, an adaptive control law with two inputs is proposed based on Lyapunov stability theory. Secondly, two other control schemes are obtained based on a further analysis of the proposed adaptive control law. Finally, numerical simulations are presented to validate the effectiveness and correctness of these results.

**Keywords:** chaos synchronization, hyperchaotic liu system, adaptive control, barlat's lemma, lyapunov stability theory

## 1 INTRODUCTION

Since it was introduced by Pecora and Carroll [1] in 1990, the synchronization of chaotic systems has attracted increasing attention due to its possible applications in secure communication [2–4], biomedical Engineering [5], information science [6], chemical reactions [7, 8], etc. Then a wide variety of control methods of chaos synchronization have been studied such as the linear feedback control [9, 10], the sliding mode control [11], the adaptive control [12–16], the backstepping control [17, 18] and so on.

A hyperchaotic system is a chaotic system with at least two positive Lyapunov exponents which improves the security by generating more complex dynamics and so hyperchaotic systems have much more applications than low-dimension chaotic systems in the areas such as secure communication and image encryption, etc. Therefore in the past three decades, an increasing interest has been devoted to the study of chaos synchronization for hyperchaotic systems and plenty of research works can be found in literature [13, 19–25]. However, there are few studies about the adaptive synchronization of hyperchaotic Liu system that was introduced in [26]. A difficulty for this problem is that it is not evident to construct a suitable Lyapunov function to prove the stability of the error dynamics because of the special complex structure of hyperchaotic Liu system, since such a Lyapunov function depends on how to choose the feedback gain of controller in the slave system. Regarding those gains as unknown parameters, the use of adaptive control concept to estimate those unknown gains may be helpful to solve this problem. Therefore, this paper investigates the adaptive control design for the synchronization of hyperchaotic Liu system. First, an adaptive control law with two inputs is proposed and moreover, by introducing a suitable Lyapunov function, the main result is proved based on Lyapunov stability theory and Barbalat's lemma. In addition, two special cases of this adaptive control scheme is further analyzed so that two other control laws are derived for the synchronization of hyperchaotic Liu system.

This paper is organized as follows. **Section 2** formulates the problem. In **Section 3**, an adaptive control law for the synchronization of hyperchaotic Liu systems is proposed. In **Section 4**, two other control laws are derived based on further analysis of the proposed adaptive control law. Numerical simulations are given in **Section 5**. We conclude this paper in **Section 6**.

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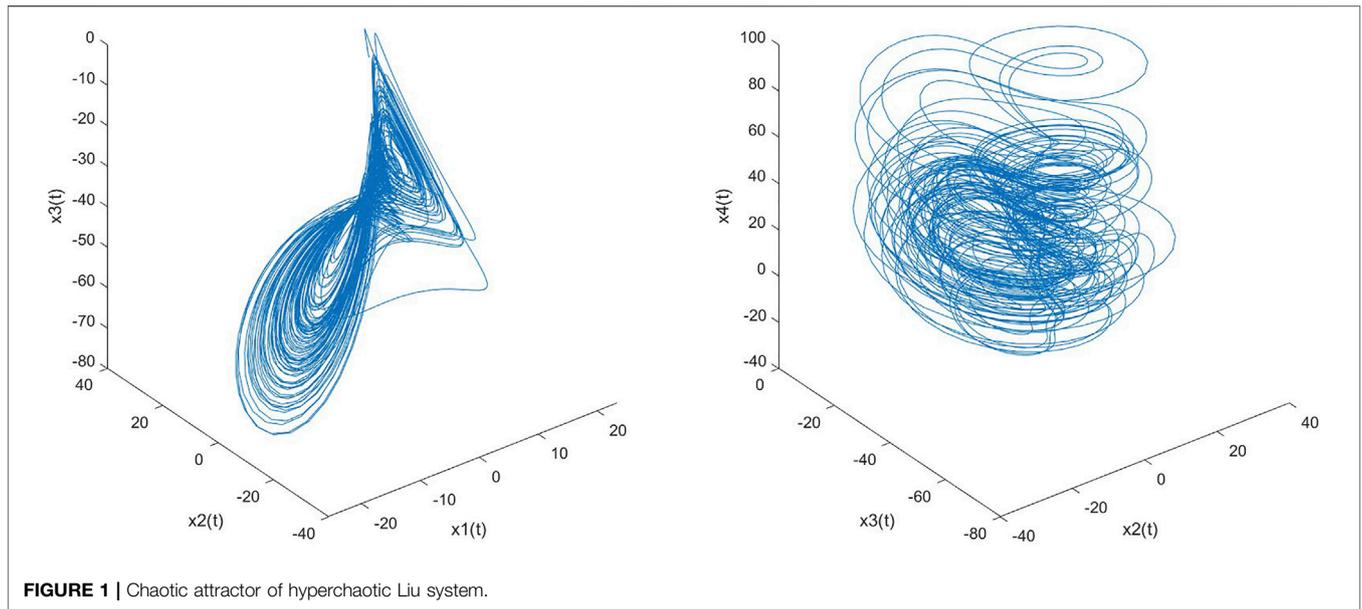
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## 2 SYSTEM DESCRIPTION AND PROBLEM FORMULATION

### 2.1 System Description

The hyperchaotic Liu system was proposed firstly in [26] and can be described by the following differential equations

$$\begin{cases} \dot{x}_1 = a(x_2 - x_1), \\ \dot{x}_2 = bx_1 - x_4 + x_1x_3, \\ \dot{x}_3 = -cx_3 + x_4 - x_1x_2, \\ \dot{x}_4 = dx_1 + x_2, \end{cases} \quad (1)$$

where  $a, b, c, d$  are positive real constants and  $x = (x_1, x_2, x_3, x_4)^T \in \mathbb{R}^4$  denotes the state vector. System (Eq. 1) was proved to exhibit hyperchaotic behavior when  $a = 10, b = 35, c = 1.4, d = 5$ . The projections of the chaotic attractor onto the  $(x_1, x_2, x_3)$  and  $(x_2, x_3, x_4)$  spaces are shown in Figure 1. Moreover, the state trajectories  $x_1(t), x_2(t), x_3(t), x_4(t)$  are globally bounded for all  $t \geq 0$  and hence, there exist four positive real constants  $M_1, M_2, M_3, M_4$  such that  $|x_1(t)| \leq M_1, |x_2(t)| \leq M_2, |x_3(t)| \leq M_3, |x_4(t)| \leq M_4$  hold for all  $t \geq 0$ .

### 2.2 Problem Formulation

Let system (Eq. 1) be the master system and the corresponding slave system is described by the following equations:

$$\begin{cases} \dot{y}_1 = a(y_2 - y_1) + u_1, \\ \dot{y}_2 = by_1 - y_4 + y_1y_3 + u_2, \\ \dot{y}_3 = -cy_3 + y_4 - y_1y_2 + u_3, \\ \dot{y}_4 = dy_1 + y_2 + u_4, \end{cases} \quad (2)$$

where  $y = (y_1, y_2, y_3, y_4)^T \in \mathbb{R}^4$  denotes the state vector and  $u = (u_1, u_2, u_3, u_4)^T$  is the control vector to be designed. Define the error states as  $e_i = y_i - x_i$ , for  $i = 1, 2, 3, 4$ , and thus the error dynamics can be described in the following form:

$$\begin{cases} \dot{e}_1 = a(e_2 - e_1) + u_1, \\ \dot{e}_2 = be_1 - e_4 + y_1e_3 + x_3e_1 + u_2, \\ \dot{e}_3 = -ce_3 + e_4 - y_1e_2 - x_2e_1 + u_3, \\ \dot{e}_4 = de_1 + e_2 + u_4. \end{cases} \quad (3)$$

It is evident that the synchronization between (Eqs 1, 2) can be easily achieved if we can use all measurements  $x_i$  of (Eq. 1) to active all controller  $u_i$  in (Eq. 2). But the interesting question is: can we use less information of (Eq. 1) to active only some of the controllers  $u_i$  in (Eq. 2)? The positive answer of this question will be more appulsive since less sensors are needed for (Eq. 1) to measure only the necessary states, and less energy will be consumed to actuate some necessary  $u_i$  in (Eq. 2). Consequently, the purpose of this paper is, by using less measurement from (Eq. 1) and activating less actuator for (Eq. 2), to design an adaptive control scheme in order to achieve global chaos synchronization of two identical hyperchaotic Liu systems, i.e., for any initial conditions  $y(0) \neq x(0)$ , we have

$$\lim_{t \rightarrow \infty} \|e(t)\| = \lim_{t \rightarrow \infty} \|y(t, y(0)) - x(t, x(0))\| = 0,$$

where  $e(t) = (e_1(t), e_2(t), e_3(t), e_4(t))^T$ .

## 3 ADAPTIVE SYNCHRONIZATION OF HYPERCHAOTIC LIU SYSTEM

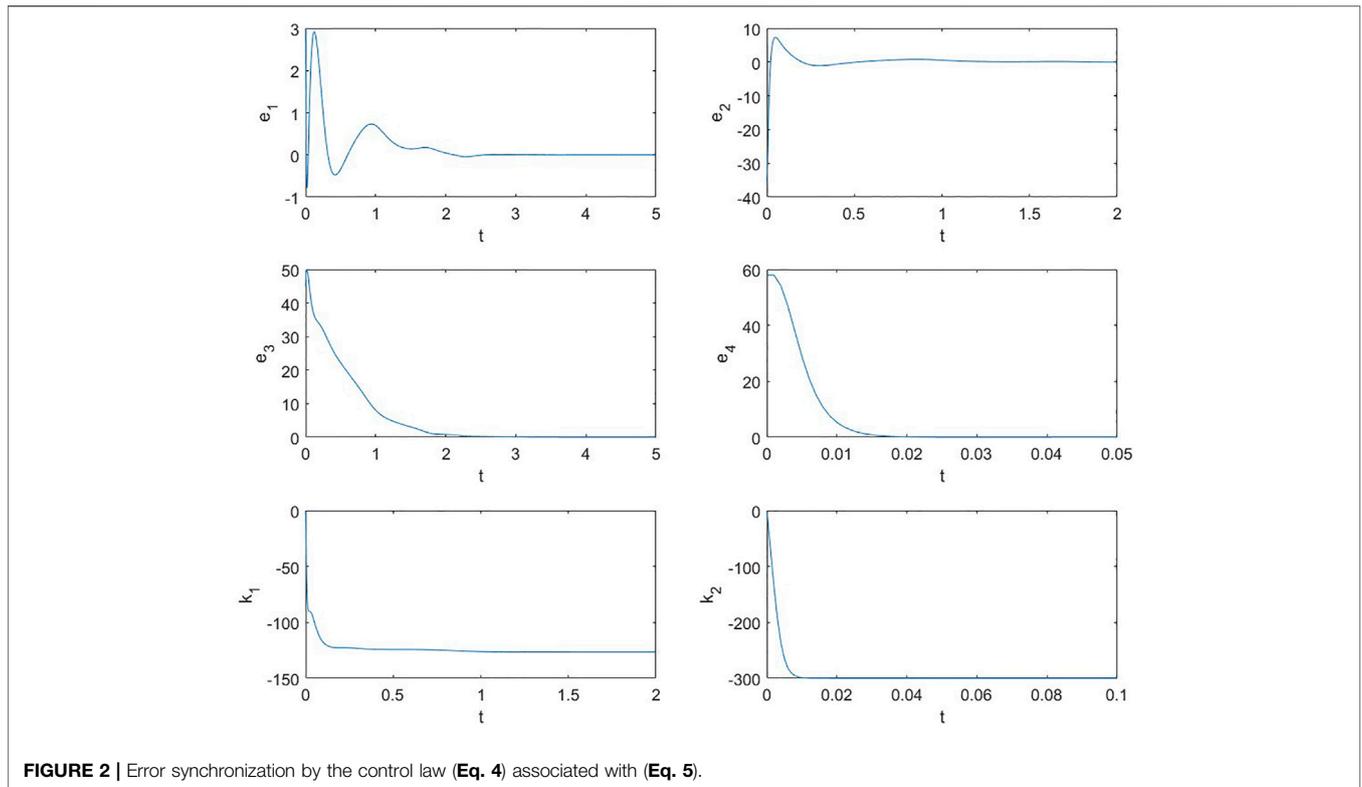
In the following, we will present our result by activating only two controllers ( $u_2$  and  $u_4$ ) in (Eq. 2) with only two measurements ( $x_2$  and  $x_4$ ) of (Eq. 1) to solve the global synchronization problem.

**Theorem 3.1.** The master system (Eq. 1) and the slave system (Eq. 2) can be global synchronized by the following controllers

$$u_1 = 0, \quad u_2 = k_1e_2, \quad u_3 = 0, \quad u_4 = k_2e_4, \quad (4)$$

where  $k_1$  and  $k_2$  denote the feedback gains which are updated by the following adaptive laws

$$\dot{k}_1 = -\gamma_1 e_2^2, \quad \dot{k}_2 = -\gamma_2 e_4^2, \quad (5)$$



**FIGURE 2 |** Error synchronization by the control law (Eq. 4) associated with (Eq. 5).

with  $\gamma_1$  and  $\gamma_2$  being arbitrary positive constants.

Proof. Under the adaptive control laws (Eqs 4, 5), the error dynamics (Eq. 3) reads

$$\begin{cases} \dot{e}_1 = a(e_2 - e_1), \\ \dot{e}_2 = be_1 - e_4 + y_1e_3 + x_3e_1 + k_1e_2, \\ \dot{e}_3 = -ce_3 + e_4 - y_1e_2 - x_2e_1, \\ \dot{e}_4 = de_1 + e_2 + k_2e_4, \\ \dot{k}_1 = -\gamma_1e_2^2, \\ \dot{k}_2 = -\gamma_2e_4^2. \end{cases} \quad (6)$$

Consider the following Lyapunov function

$$V = \frac{1}{2}(\rho e_1^2 + e_2^2 + e_3^2 + e_4^2) + \frac{1}{2\gamma_1}(k_1 + L_1)^2 + \frac{1}{2\gamma_2}(k_2 + L_2)^2, \quad (7)$$

where  $L_1, L_2$  and  $\rho$  are positive constants which will be determined later. Calculating the differentiation of (Eq. 7) with respect to time  $t$  along trajectories of system (Eq. 6), we obtain

$$\begin{aligned} \dot{V} &= \rho e_1 \dot{e}_1 + e_2 \dot{e}_2 + e_3 \dot{e}_3 + e_4 \dot{e}_4 + \frac{1}{\gamma_1}(k_1 + L_1)\dot{k}_1 + \frac{1}{\gamma_2}(k_2 + L_2)\dot{k}_2 \\ &= \rho a e_1(e_2 - e_1) + e_2(be_1 - e_4 + y_1e_3 + x_3e_1 - L_1e_2) \\ &\quad + e_3(-ce_3 + e_4 - y_1e_2 - x_2e_1) + e_4(de_1 + e_2 - L_2e_4) \\ &= -\rho a e_1^2 - L_1 e_2^2 - ce_3^2 - L_2 e_4^2 + (\rho a + b + x_3)e_1 e_2 \\ &\quad - x_2 e_1 e_3 + de_1 e_4 + e_3 e_4 \\ &\leq -\left(\rho a - \frac{d}{2}\right)e_1^2 - L_1 e_2^2 - \left(c - \frac{1}{2}\right)e_3^2 - \left(L_2 - \frac{d+1}{2}\right)e_4^2 \\ &\quad + (\rho a + b + M_3)|e_1 e_2| + M_2|e_1 e_3| \\ &= -|e|^T P_1 |e|, \end{aligned} \quad (8)$$

where  $|e| = (|e_1|, |e_2|, |e_3|, |e_4|)^T$ ,  $M_2$  and  $M_3$  represent the boundedness of  $x_2$  and  $x_3$ , and  $P_1$  is a symmetric matrix of the following form

$$P_1 = \begin{pmatrix} \rho a - \frac{d}{2} & \frac{\rho a + b + M_3}{2} & -\frac{M_2}{2} & 0 \\ -\frac{\rho a + b + M_3}{2} & L_1 & 0 & 0 \\ -\frac{M_2}{2} & 0 & c - \frac{1}{2} & 0 \\ 0 & 0 & 0 & L_2 - \frac{d+1}{2} \end{pmatrix}. \quad (9)$$

Clearly, if the symmetric matrix  $P_1$  is positive definite, then we have  $\dot{V} \leq 0$ . Moreover, it is well known that  $P_1$  is positive definite if and only if  $\Delta_k > 0$ , for  $1 \leq k \leq 4$ , where  $\Delta_k$  denotes the Leading Principle Minor of order  $k$  of  $P_1$ . A straightforward calculation gives

$$\begin{aligned} \Delta_1 &= \rho a - \frac{d}{2}, \\ \Delta_2 &= L_1 \Delta_1 - \frac{(\rho a + b + M_3)^2}{4}, \\ \Delta_3 &= \left(c - \frac{1}{2}\right) \Delta_2 - L_1 \frac{M_2^2}{4}, \\ \Delta_4 &= \left(L_2 - \frac{d+1}{2}\right) \Delta_3. \end{aligned} \quad (10)$$

From the condition  $\Delta_k > 0$ , for  $1 \leq k \leq 4$ , and (Eq. 10), it is easy to obtain

$$L_1 > \max \left\{ \frac{(\rho a + b + M_3)^2}{4a\rho - 2d}, \frac{(c - 1/2)(\rho a + b + M_3)^2}{(2c - 1)(2a\rho - d) - M_2^2} \right\},$$

$$L_2 > \frac{d + 1}{2}, \quad \rho > \frac{1}{2a} \left( \frac{M_2^2}{2c - 1} + d \right).$$

Therefore, there always exist positive  $L_1, L_2$  and  $\rho$  satisfying the above conditions so that  $\dot{V} \leq 0$  and thus  $V$  is positive and decrescent. It follows that the equilibrium point  $(e_1 = 0, e_2 = 0, e_3 = 0, e_4 = 0, k_1 = k_1^*, k_2 = k_2^*)$  of systems (Eq. 6) is uniformly stable which implies, from (Eq. 6), that  $\dot{e}_1(t), \dot{e}_2(t), \dot{e}_3(t), \dot{e}_4(t)$  are also uniformly stable. Moreover, it is easy to conclude from (Eq. 8) that the error states  $e_1(t), e_2(t), e_3(t), e_4(t)$  are quadratically integrable function, i.e.,  $e_1(t), e_2(t), e_3(t), e_4(t) \in L_2$ . Therefore, according to Barbalat's lemma, given any initial conditions, we have always  $e_1(t) \rightarrow 0, e_2(t) \rightarrow 0, e_3(t) \rightarrow 0, e_4(t) \rightarrow 0$  ( $t \rightarrow \infty$ ) and  $k_1 \rightarrow k_1^*, k_2 \rightarrow k_2^*$  ( $t \rightarrow \infty$ ), which imply that the master system (Eq. 1) and slave system (Eq. 2) are globally asymptotically synchronized under the adaptive control law (Eq. 4) associated with (Eq. 5).

Remark 3.1. In the Lyapunov function (Eq. 7), a positive constant  $\rho$  has to be introduced to guarantee the positive definite of the symmetric matrix  $P_1$ .

Remark 3.2. The feedback gains  $k_1$  and  $k_2$  will converge to two constants  $k_1^*$  and  $k_2^*$ , respectively, which depend on not only the initial condition  $k_1(0)$  and  $k_2(0)$ , but also the value of  $\gamma_1$  and  $\gamma_2$ .

### 4 FURTHER ANALYSIS OF THEOREM 3.1

In Theorem 3.1, the feedback gains  $k_1$  and  $k_2$  are defined to be updated by the adaptive laws (Eq. 5), i.e.,

$$\dot{k}_1 = -\gamma_1 e_2^2, \quad \dot{k}_2 = -\gamma_2 e_4^2,$$

where  $\gamma_1$  and  $\gamma_2$  are arbitrary positive constants. Two interesting questions arise:

- If  $\gamma_1 = \gamma_2 = 0$  that implies  $\dot{k}_1 = \dot{k}_2 = 0$ , i.e.,  $k_1$  and  $k_2$  are both constants equal to their initial values, then is the control law (Eq. 4) still valid?
- If the feedback gains  $k_1$  and  $k_2$  are equal (for the sake of implementation simplicity), i.e.,  $k_1 = k_2$ , how to modify the adaptive law (Eq. 5) such that the chaos synchronization can still be achieved?

In the following, we will answer the above two questions and derive two other special control laws for the chaos synchronization of hyperchaotic Liu system.

#### 4.1 Case 1: $k_1$ and $k_2$ Are Constants

Proposition 4.1. The master system (Eq. 1) and the slave system (Eq. 2) can be global synchronized by the following linear feedback controller

$$u_1 = 0, \quad u_2 = k_1 e_2, \quad u_3 = 0, \quad u_4 = k_2 e_4, \quad (11)$$

where the feedback gains  $k_1, k_2$  satisfy

$$k_1 < \min \left\{ -\frac{(\rho a + b + M_3)^2}{4a\rho - 2d}, -\frac{(c - 1/2)(\rho a + b + M_3)^2}{(2c - 1)(2a\rho - d) - M_2^2} \right\},$$

$$k_2 < -\frac{d + 1}{2},$$

with

$$\rho > \frac{1}{2a} \left( \frac{M_2^2}{2c - 1} + d \right). \quad (13)$$

Proof. The proof of the above result is quite similar to that of Theorem 3.1. Consider the following Lyapunov function

$$V = \frac{1}{2} (\rho e_1^2 + e_2^2 + e_3^2 + e_4^2), \quad (14)$$

where  $\rho$  is a positive constant which will be determined later. Taking the differentiation of (Eq. 14) with respect to time  $t$  and following the similar procedure in the proof of Theorem 3.1, we obtain

$$\dot{V} \leq -(|e_1|, |e_2|, |e_3|, |e_4|) P_2 (|e_1|, |e_2|, |e_3|, |e_4|)^T, \quad (15)$$

where

$$P_2 = \begin{pmatrix} \rho a - \frac{d}{2} & -\frac{\rho a + b + M_3}{2} & -\frac{M_2}{2} & 0 \\ -\frac{\rho a + b + M_3}{2} & -k_1 & 0 & 0 \\ -\frac{M_2}{2} & 0 & c - \frac{1}{2} & 0 \\ 0 & 0 & 0 & -k_2 - \frac{d + 1}{2} \end{pmatrix}. \quad (16)$$

It is easy to see that the symmetric matrix  $P_2$  has the same structure with  $P_1$ , given by (Eq. 9), and the only difference is that  $L_1$  and  $L_2$  are replaced by  $-k_1$  and  $-k_2$ , respectively. Therefore, by a similar procedure of the proof of Theorem 3.1, we can obtain that  $P_2$  is positive definite if and only if the following conditions are satisfied

$$k_1 < \min \left\{ -\frac{(\rho a + b + M_3)^2}{4a\rho - 2d}, -\frac{(c - 1/2)(\rho a + b + M_3)^2}{(2c - 1)(2a\rho - d) - M_2^2} \right\}$$

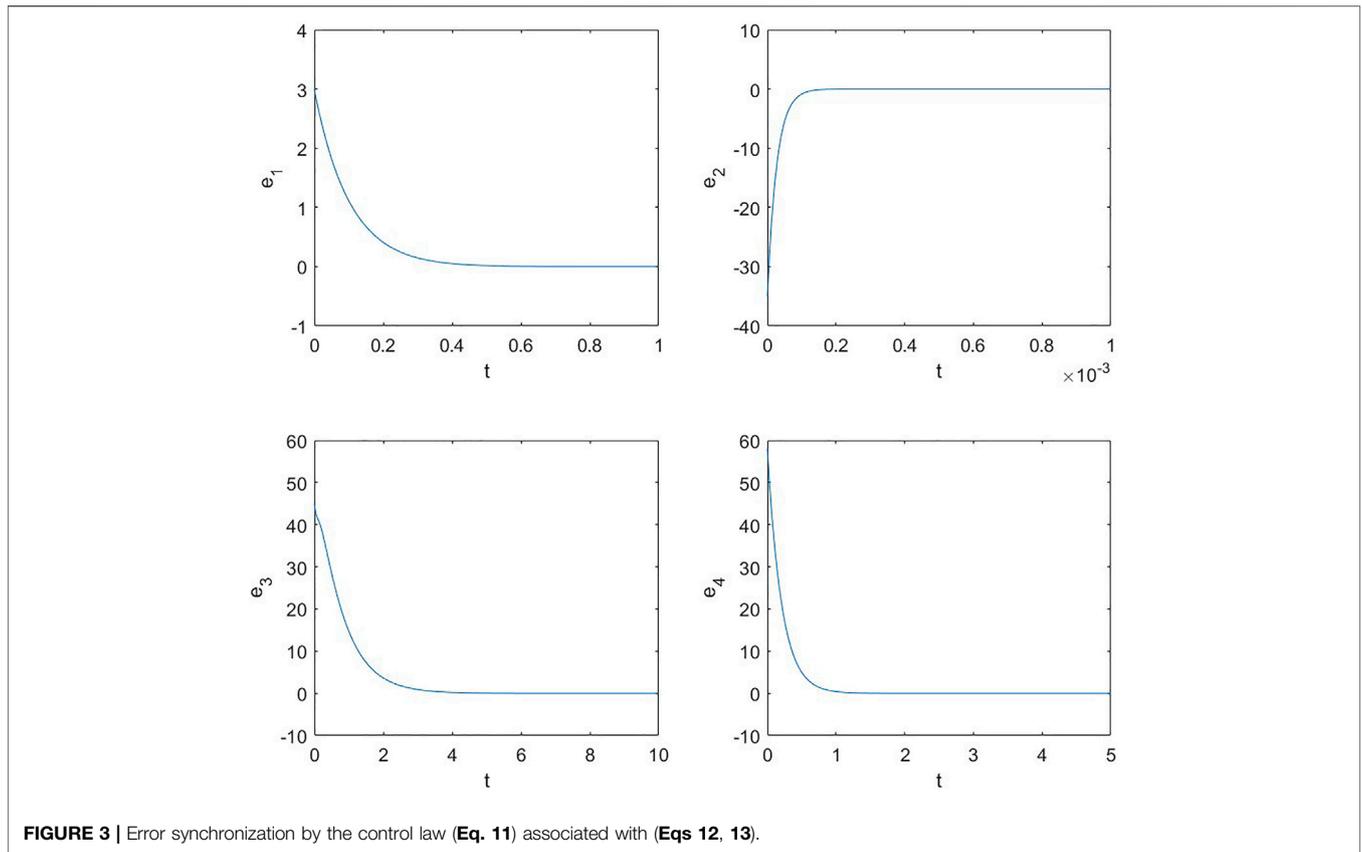
$$k_2 < -\frac{d + 1}{2}, \quad \text{and} \quad \rho > \frac{1}{2a} \left( \frac{M_2^2}{2c - 1} + d \right).$$

Therefore, there always exist  $k_1, k_2$  and  $\rho$  satisfying the above conditions so that  $\dot{V} \leq 0$  and thus  $V$  is positive and decrescent. Clearly  $\dot{V} = 0$  if and only if  $e_i = 0, 0 \leq i \leq 4$  which means the set  $R = \{e \in \mathbb{R}^4: \dot{V} = 0\}$  contains no other trajectories except  $\{e_1 = 0, e_2 = 0, e_3 = 0, e_4 = 0\}$ . Therefore, by the LaSalle invariance principle, starting with arbitrary initial values,  $e = 0$  is asymptotically stable which implies that the chaos synchronization of hyperchaotic Liu system is achieved by the control law (Eqs 11, 12).

#### 4.2 Case 2: $k_1$ and $k_2$ Are Equal

Proposition 4.2. The master system (Eq. 1) and the slave system (Eq. 2) can be synchronized by the following adaptive controller

$$u_1 = 0, \quad u_2 = k e_2, \quad u_3 = 0, \quad u_4 = k e_4, \quad (17)$$



**FIGURE 3 |** Error synchronization by the control law (Eq. 11) associated with (Eqs 12, 13).

where  $k$  denotes the feedback gain which is updated by the following adaptive law

$$\dot{k} = -\gamma(e_2^2 + e_4^2), \tag{18}$$

with  $\gamma$  being an arbitrary positive constant.

Proof. Following the same procedure as that in the proof of Theorem 3.1, by considering the following Lyapunov function

$$V = \frac{1}{2}(\rho e_1^2 + e_2^2 + e_3^2 + e_4^2) + \frac{1}{2\gamma}(k + L)^2 \tag{19}$$

where  $\rho$  and  $L$  are positive constants which will be determined later, we obtain

$$\dot{V} \leq -(|e_1|, |e_2|, |e_3|, |e_4|)P_3(|e_1|, |e_2|, |e_3|, |e_4|)^T, \tag{20}$$

with

$$P_3 = \begin{pmatrix} \rho a - \frac{d}{2} & -\frac{\rho a + b + M_3}{2} & -\frac{M_2}{2} & 0 \\ -\frac{\rho a + b + M_3}{2} & L & 0 & 0 \\ -\frac{M_2}{2} & 0 & c - \frac{1}{2} & 0 \\ 0 & 0 & 0 & L - \frac{d + 1}{2} \end{pmatrix}.$$

Clearly, the symmetric matrix  $P_3$  has the same structure with  $P_1$ , given by (Eq. 9), in which  $L_1 = L_2 = L$ . Thus, by a similar procedure

of the proof of Theorem 3.1, we can obtain that  $P_3$  is positive definite if and only if the following conditions are satisfied

$$L > \max \left\{ \frac{d + 1}{2}, \frac{(\rho a + b + M_3)^2}{4\rho - 2d}, \frac{(c - 1/2)(\rho a + b + M_3)^2}{(2c - 1)(2\rho - d) - M_2^2} \right\}$$

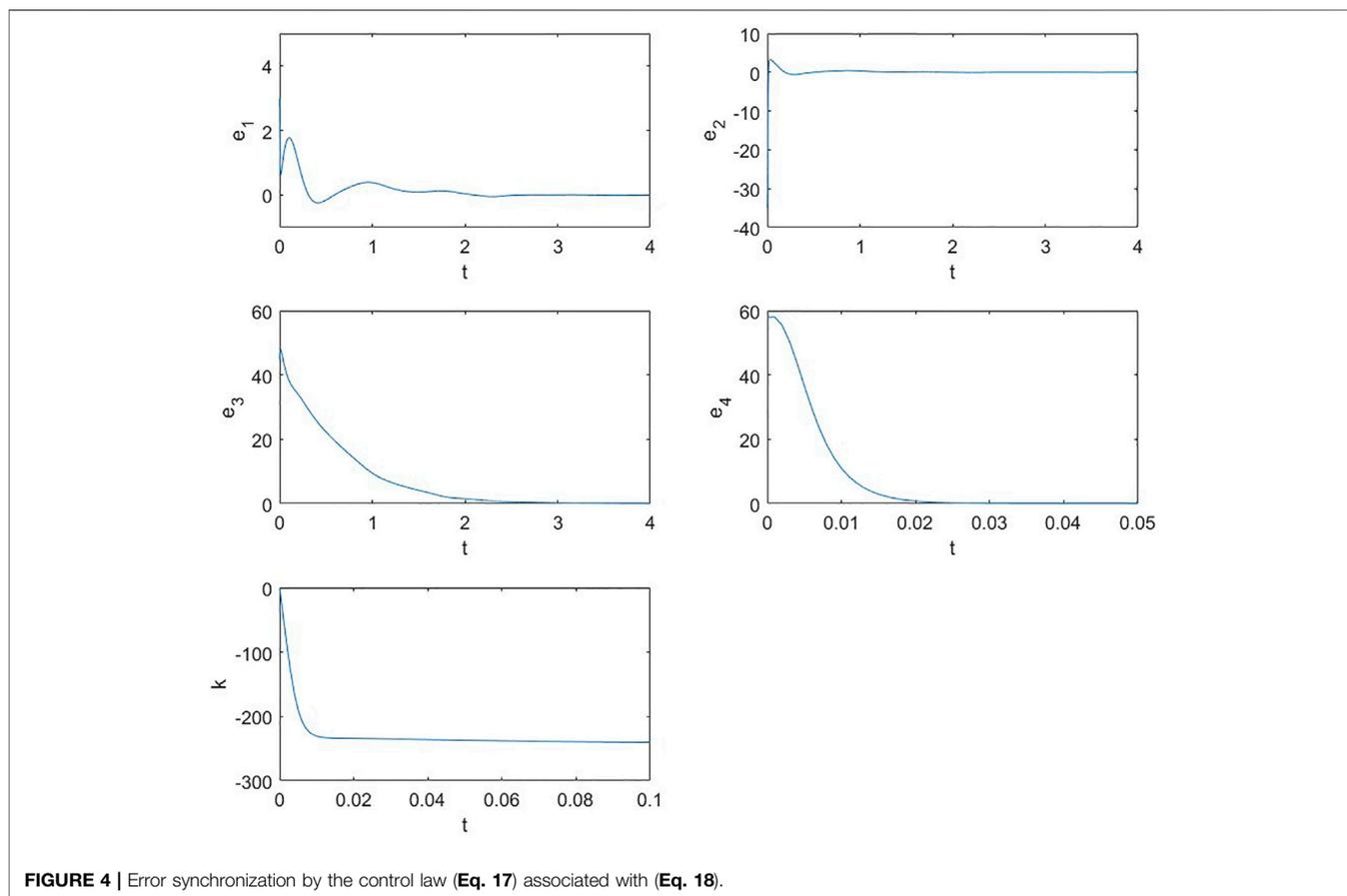
$$\rho > \frac{1}{2a} \left( \frac{M_2^2}{2c - 1} + d \right).$$

The rest of the proof is exactly the same as that in the proof of Theorem 3.1.

### 5 NUMERICAL SIMULATIONS

In this section, numerical simulations by MATLAB will be given to validate the correctness and effectiveness of the proposed controller designs. Fourth order Runge-Kutta method is applied to approximate the solution of differential equations with a small chosen fixed time step size. The system parameters are set to  $a = 10, b = 35, c = 1.4, d = 5$  so that the Hyperchaotic Liu system exhibits chaotic behavior when no control is applied.

In order to compare the three control laws proposed in Theorem 3.1, Proposition 4.1 and Proposition 4.2, we will use the same initial conditions  $x(0)$  and  $y(0)$  for all the three numerical simulations. The initial conditions of the master system are chosen as  $x_1(0) = 15, x_2(0) = 22, x_3(0) = -46$  and



**FIGURE 4** | Error synchronization by the control law (Eq. 17) associated with (Eq. 18).

$x_4(0) = -21$  while the initial conditions of the slave system are chosen as  $y_1(0) = 18$ ,  $y_2(0) = -13$ ,  $y_3(0) = -1$  and  $y_4(0) = 37$ .

First, consider the adaptive control law (Eq. 4) associated with (Eq. 5) proposed in Theorem 3.1 and select  $\gamma_1 = 10$ ,  $\gamma_2 = 20$  and the initial conditions of the feedback gain by  $k_1(0) = k_2(0) = 1$ . Figure 2 shows that the error states are asymptotically stable to zero while the control gains  $k_1$  and  $k_2$  tend to two negative constants, respectively, as  $t$  tends to infinity.

Second, consider the control law proposed in Proposition 4.2 and select upper bounds of  $M_2$  and  $M_3$  by  $M_2 = 52$ ,  $M_3 = 82$  (according to the bounded simulation results depicted in Figure 1). From conditions (Eqs 12, 13), a direct calculation gives that  $k_1 < -30\,097$  and  $k_2 < -3$  and we choose  $k_1 = -31\,000$  and  $k_2 = -5$ . Figure 3 shows that the error states asymptotically converge to zero, as  $t$  tends to infinity.

Finally, consider the adaptive control law (Eq. 17) associated with (Eq. 18) proposed in Theorem 3.1 and select  $\gamma = 10$  and  $k(0) = 1$ . Figure 4 shows that the error system is asymptotically stable to zero and the control gain  $k$  tends to a negative constant as  $t$  tends to infinity.

## 6 CONCLUSION

This paper has addressed the adaptive synchronization problems of hyperchaotic Liu systems. An adaptive control scheme has been

proposed for the asymptotically synchronization of two identical hyperchaotic Liu systems. This result was proved according to Lyapunov stability theory and Barbalat's lemma by constructing a suitable Lyapunov function. Moreover, through further discussions of the main result, two other control schemes have been derived. The numerical simulations verify the effectiveness and correctness of the control laws proposed in this work.

## DATA AVAILABILITY STATEMENT

The original contributions presented in the study are included in the article/Supplementary Material, further inquiries can be directed to the corresponding author.

## AUTHOR CONTRIBUTIONS

SL: Conceptualization, Methodology, Project administration, Funding acquisition. YW: Validation, Investigation, Writing–Original Draft. GZ: Supervision, Writing–Review and Editing.

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